

WD3136E

24V DC/DC Step-Up Regulator

Descriptions

The WD3136E is a high efficiency, high power, peak current mode step-up converter integrated 32V power MOSFET with 3A (Typ.) current limit. It could support up to maximum 24V output.

The boost converter WD3136E runs in Pulse-Width Modulation (PWM) mode at 1MHz fixed switching frequency to reduce output ripple and improve conversion efficiency. It allows for the use of small external components. At light load, the converter enters Skipping Mode to maintain a high efficiency over a wide load current range. The build-in soft start circuitry minimizes the inrush current at start-up.

The WD3136E is available in SOT-23-6L package. Standard product is Pb-free and Halogen-free.

Features

Input Voltage Range : 2.4V~20V
Reference Voltage : 600mV
Switching Frequency : 1MHz (Typ.)
Efficiency : Up to 94%

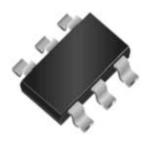
Internal Compensation

Current Limit : 3A (D=50%)
Low R_{DS(ON)} : 100mΩ
Automatic Pulse Skipping Modulation
Package : SOT-23-6L

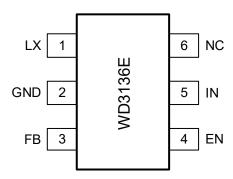
Applications

- Cell Phone and Smart Phone
- PDA, PMP, MP3
- Digital Camera

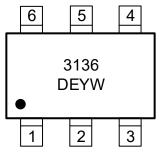
Http//:www.ovt.com



SOT-23-6L



Pin Configuration (Top View)



3136 = Device code

DE= Special code

Y = Year code

W = Week code

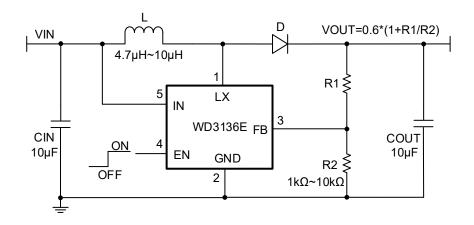
Marking

Order Information

Device	Package	Shipping
WD3136E-6/TR	SOT-23-6L	3000/Reel & Tape



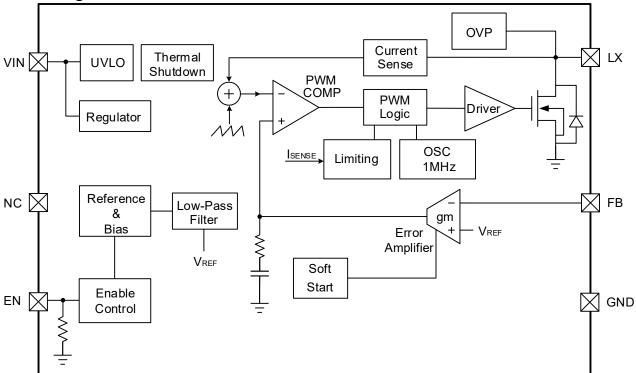
Typical Applications



Pin Descriptions

Symbol	Pin No.	Descriptions		
LX	1	Switch Node		
GND	2	Ground		
FB	3	Feedback		
EN	4	Enable		
IN	5	Power Supply		
NC	6	No Connection		





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Absolute Maximum Ratings

Parameter	Symbol	Value	Unit
VIN Pin Voltage Range	Vin	-0.3~28	V
EN Pin Voltage Range	V_{EN}	-0.3~28	V
LX Pin Voltage Range (DC)	V_{LX}	-0.3~32	V
FB Pin Voltage Range	V _{FB}	-0.3~6	V
Junction To Ambient Thermal Resistance – SOT-23-6L (Note 1)	RθJA	90	°C/W
Operating Junction Temperature	ТJ	-40~150	°C
Lead Temperature(Soldering, 10s)	T∟	260	°C
Storage Temperature	T _{stg}	-55~150	°C

These are stress ratings only. Stresses exceeding the range specified under "Absolute Maximum Ratings" may cause substantial damage to the device. Functional operation of this device at other conditions beyond those listed in the specification is not implied and prolonged exposure to extreme conditions may affect device reliability.

Note 1: Thermal Resistance (θ_{JA}) is measured with the component mounted on 1.5inch x 1.5inch, 2layers, FR4 test board with 1.0inch x 1.0inch copper area of 2oz in top layer, and in still air condition.

ESD Ratings

Parameter	Symbol	Value	Unit
Clastrastatia Discharge	НВМ	4000	V
Electrostatic Discharge	CDM	2000	V

Recommended Operating Conditions

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Parameter	Symbol	Value	Unit
Input Voltage Range	Vin	2.4~20	V
Output Voltage Range	Vouт	V _{IN} ~24	V
Inductor	L	4.7~10	μH
Input Capacitor	C _{IN}	10 (Typ.)	μF
Output Capacitor	Соит	10 (Typ.)	μF
Resistance Between FB Pin And GND	R ₂	1~10	kΩ
Operating Ambient Temperature	T _A	-40~85	°C
Operating Junction Temperature	TJ	-40~125	°C

These values are recommended values that have been successfully tested in several applications. Other values may be acceptable in other applications but should be fully tested by the user.



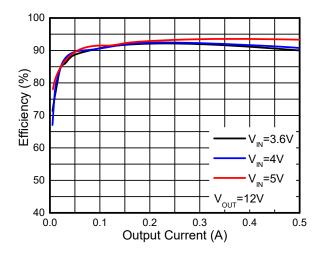
Electronics Characteristics

(Ta=25 °C, V_{IN}=3V, L=4.7μH, C_{IN}=C_{OUT}=10μF, unless otherwise noted)

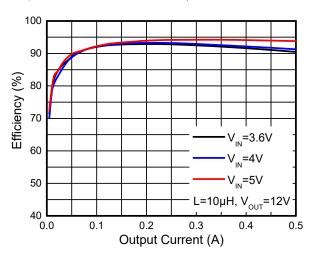
Parameter	Symbol	Test Condition	Min	Тур	Max	Units
Basic Operation				•		
Operation Voltage Range	V _{IN}		2.4		20	V
Under Voltage Lockout	V _{UVLO}	V _{IN} Rising		2.2	2.39	V
UVLO Hysteresis	V _{HYS}			0.1		V
Quiescent Current	ΙQ	No Switching V _{FB} =0.7V		0.25		mA
Supply Current	Is	Switching at no load		0.5		mA
Shutdown Current	IsD	V _{EN} <0.4V		0.05	1	μΑ
Enable Control						
Enable Low Voltage	V _{ENL}				0.4	V
Enable High Voltage	V _{ENH}		1.5			V
EN Pull-down Resistance	Ren			1.8		МΩ
EN Shutdown Delay	tshon, en			0.12		ms
Voltage And Current Control	•		•			•
Operation Frequency	fosc		8.0	1	1.2	MHz
Maximum Duty Cycle	D _{MAX}			91		%
Feedback Reference	V _{FB}		0.588	0.6	0.612	V
Feedback Input Bias Current	I _{FB}			0.1	1	μA
Power Switch						
On Resistance	Ron	I _{LX} =1A		0.1		Ω
Main Switch Leakage Current	I _{LN_MS}	V _{LX} =32V, Disable		0.1	1	μA
Current Limit	I _{LIM}	D=50%		3		Α
Protection						
Thermal Shutdown	T _{SD}			160		°C
T _{SD} Hysteresis	T _{SD-HYS}			30		°C

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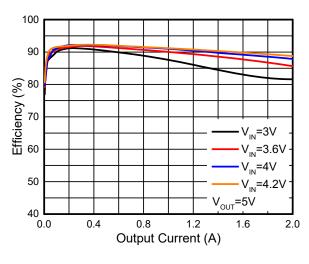
Typical Characteristics (Ta=25 °C, L=4.7μH, C_{IN}=C_{OUT}=10μF, unless otherwise noted)



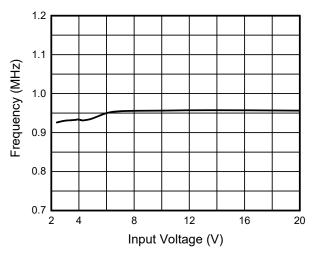
Efficiency vs. Output Current



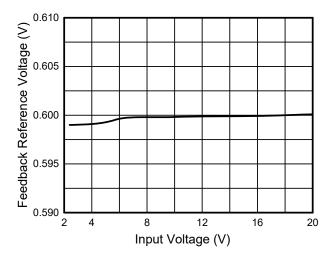
Efficiency vs. Output Current



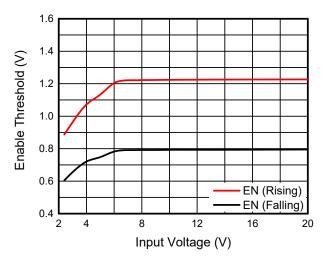
Efficiency vs. Output Current



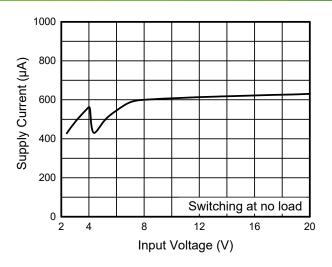
Frequency vs. Input Voltage



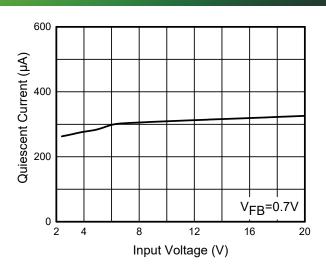
Feedback Reference Voltage vs. Input Voltage



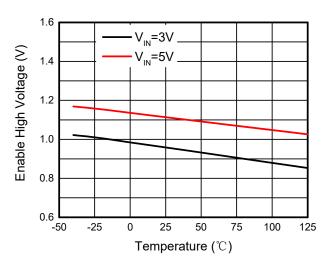
Enable Threshold vs. Input Voltage



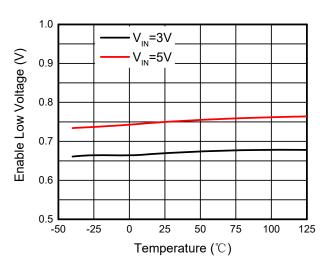
Supply Current vs. Input Voltage



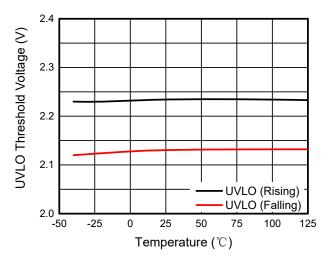
Quiescent Current vs. Input Voltage



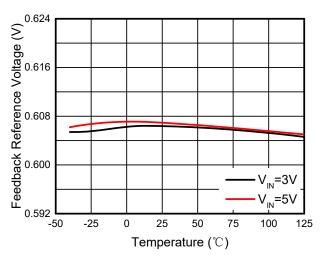
Enable High Voltage vs. Temperature



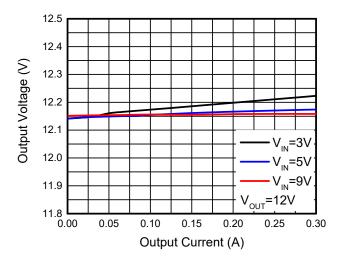
Enable Low Voltage vs. Temperature



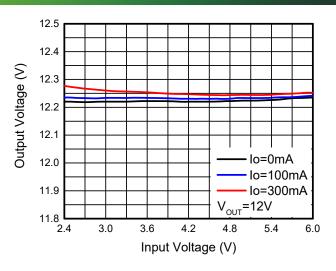
UVLO Threshold Voltage vs. Temperature



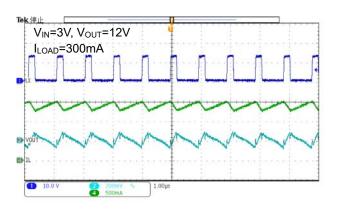
Feedback Reference Voltage vs. Temperature



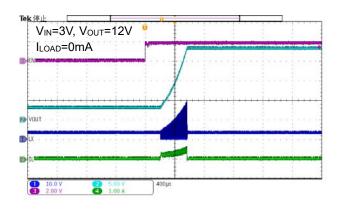




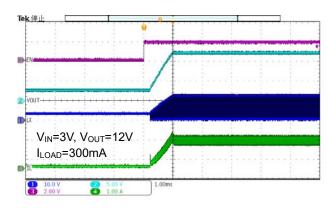
Output Voltage vs. Input Voltage



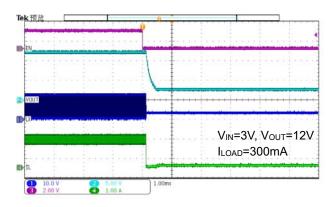
Operation Waveforms



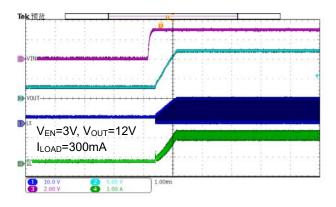
Output Open Circuit Protection

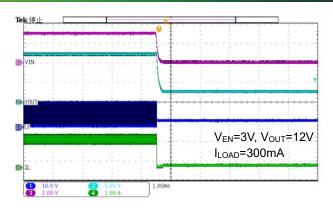


Start-Up From EN



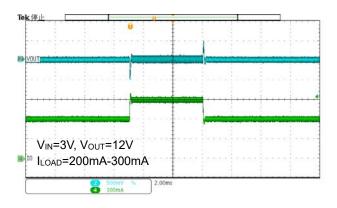
Shut-Down From EN





Start-Up From IN

Shut-Down From IN



Load Transient Response

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Operation Information

Normal Operation

The WD3136E is a high efficiency, high output voltage boost converter. The device uses a fixed frequency, peak current mode boost regulator architecture to regulate voltage at the feedback pin. The device integrates 32V/3A switch MOSFET. The operation of the WD3136E can be understood by referring to the block diagram. The beginning of each cycle turns on the Power MOSFET. A slope compensation ramp is added to output of the current sense amplifier and the result is fed into the negative input of the comparator (COMP). When this voltage goes above the output voltage of the error amplifier (EA), the Power MOSFET is turned off. The FB voltage can be regulated to the reference voltage of bandgap with EA block. The feedback loop regulates the FB pin to a low reference voltage (600mV Typ.). The WD3136E has internal soft start to limit the amount of input current at startup and to also limit the amount of overshoot on the output.

Cycle-By-Cycle Current Limit

The WD3136E uses a cycle-by-cycle current limit circuitry to limit the inductor peak current in the event of an overload condition. The current flow through inductor in charging phase is detected by a current sensing circuit. As the value comes across the current limiting threshold the power MOSFET turns off, so that the inductor will be forced to leave charging stage and enter discharging stage. Therefore, the inductor current will not increase over the current limiting threshold.

Output Open Circuit Protection

Output open circuit protection circuitry prevents IC damage as the result of output open circuit. The WD3136E monitors the voltage at the LX pin and FB pin during each switching cycle. The circuitry turns off the switch MOSFET and shuts down the IC when both of the following conditions persist for 4 switching clock cycles: (1) the LX voltage exceeds the set

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threshold voltage and (2) the FB voltage is less than 50mV. Then, the WD3136E turns off the power switch MOSFET and shuts down IC until EN or power supply is recycled to enable IC.

UVLO Protection

To avoid malfunction of the WD3136E at low input voltages, an under voltage lockout is included that disables the device, until the input voltage exceeds 2.2V (Typ.).

Shutdown Mode

Drive EN to GND to place the WD3136E in shutdown mode. In shutdown mode, the reference, control circuit, and the main switch turn off. Input current falls to smaller than 1µA during shutdown mode.

Over-Temperature-Protection (OTP)

As soon as the junction temperature (T_J) exceeds 160°C (Typ.), the WD3136E goes into thermal shutdown. In this mode, the main power MOSFET is turned off until temperature falls below 130°C (Typ.). Then the device starts switching again.





Application Information

External component selection for the application circuit depends on the load current requirements. Certain trade-offs between different performance parameters can also be made.

Output Voltage Setting

The loop of Boost structure will keep the FB pin voltage equal to the reference voltage VREF. Therefore, when R₂ connects FB pin and GND, R₁ connects OUT pin and FB pin, the output voltage is set externally by using a resistor divider, which is equal to following equation:

$$V_{out} = V_{FB} \times (1 + \frac{R_1}{R_2})$$

Where

V_{OUT} = output voltage

V_{FB} = regulated voltage of FB pin

 R_1 = resistance between OUT pin and FB pin

R₂ = resistance between FB pin and GND

The recommended value of R_2 is $1k\Omega \sim 10k\Omega$.

The output current tolerance depends on the V_{FB} accuracy and the current sensor resistor accuracy.

Boost Inductor Selection

Small inductance will make the ripple current large and too large inductance will cause poor dynamic characteristics and the slow response. The proper inductance should be selected to ensure the loop stability. For these reasons, the recommended range of inductor value for the application is $4.7\mu H \sim 10\mu H$.

Input Capacitor Selection

Connect the input capacitance from the IN pin to the reference ground plane. Input capacitance reduces the ac voltage ripple on the input rail for the boost converter. The rated voltage of capacitor depends on the applied input voltage. The recommended capacitor is a 10µF/50V, X5R or X7R ceramic capacitor.

Output Capacitor Selection

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Santa Clara, CA 95054

The output capacitor is mainly selected to meet the

requirements for the output ripple and loop stability. This ripple voltage is related to the capacitor's capacitance and its equivalent series resistance (ESR). The rated voltage of capacitor depends on the applied output voltage. The recommended capacitor is a 10µF/50V, X5R or X7R ceramic capacitor.

Diode Selection

The rectifier diode supplies current path to the inductor when the internal MOSFET is off. Use a schottky diode with low forward voltage to reduce losses. The diode should be rated for a reverse blocking voltage greater than the output voltage used. The average current rating must be greater than the maximum load current expected, and the peak current rating must be greater than the peak inductor current.

PCB Layout Considerations

- Use separated power supply trace and power ground planes from other sensitive blocks.
- 2. Locate C_{IN} as close to the IN pin as possible. And connect the C_{IN}'s GND close to IC's GND.
- Route the high current path from C_{IN}, through L to the LX pin and GND as short as possible. And keep high current traces as short and as wide as possible.
- Route the high current path from L to diode and C_{OUT} as short as possible. Connect the C_{OUT}'s GND as close to IC's GND as possible.
- Avoid routing sensitive trace near this block, especially LX node. Place a ground plane shield between the traces.
- Place the R₂ resistor as close to FB pin as possible, for the FB pin is a high impedance input pin which is susceptible to noise and high voltage spike.
- 7. Avoid routing a long OUT or FB trace parallel to other sensitive signal. Place a ground plane shield between the traces.

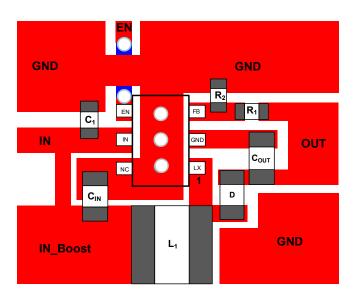


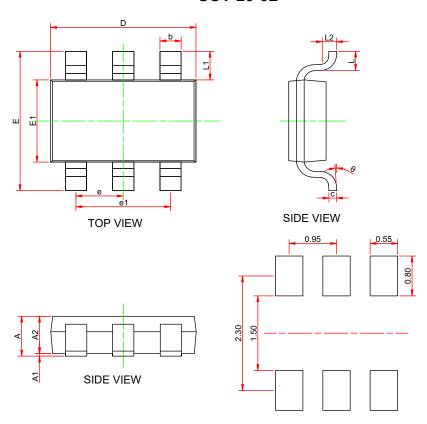
Figure 1: WD3136E PCB Layout

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PACKAGE OUTLINE DIMENSIONS

SOT-23-6L



RECOMMENDED LAND PATTERN(unit:mm)

O. wale of	Dii	Dimensions in Millimeters				
Symbol	Min.	Тур.	Max.			
А	-	-	1.25			
A1	0	-	0.15			
A2	1.00	1.10	1.20			
b	0.30	0.40	0.50			
С	0.10	-	0.21			
D	2.72	2.92	3.12			
E	2.60	2.80	3.00			
E1	1.40	1.40 1.60				
е		0.95BSC				
e1	1.80	1.90	2.00			
L	0.30	0.30 - 0.				
L1		0.59Ref				
L2		0.25Ref				
θ	0 ° - 8 °					

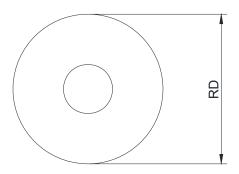
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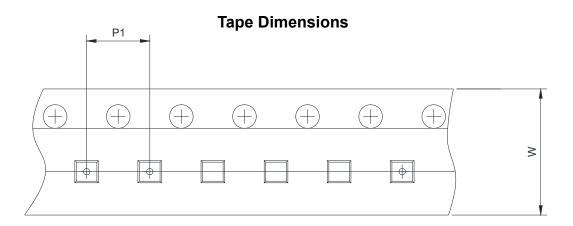
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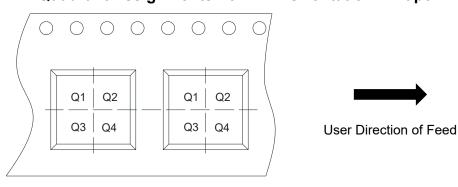
TAPE AND REEL INFORMATION

Reel Dimensions





Quadrant Assignments For PIN1 Orientation In Tape



RD	Reel Dimension	☑ 7inch	☐ 13inch		
W	Overall width of the carrier tape	☑ 8mm	☐ 12mm	☐ 16mm	
P1	Pitch between successive cavity centers	☐ 2mm	✓ 4mm	☐ 8mm	
Pin1	Pin1 Quadrant	□ Q1	□ Q2	▼ Q3	□ Q4